

## CHAPTER 2 PROPERTIES OF WAVEGUIDES

### 2.2 2-D Waveguides

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This document calculates various quantities for rectangular waveguides. For a 2-D waveguide, cutoff frequencies and guide wavelengths are calculated. Values for the cutoff frequency are calculated for  $TE_{mn}$  and  $TM_{mn}$  modes over a range of values for  $m$  and  $n$ , and a corresponding  $kz-k$  diagram is shown for the 2-D guide. You provide the following values:

- $a$  and  $b$ , the waveguide dimensions,
- $m$  and  $n$ , mode numbers,
- $f$ , the operating frequency.

#### Background

Refer to the following diagram for dimensions and orientation.

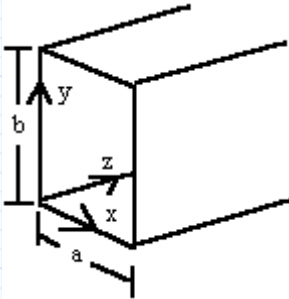


Fig. 2.2.1 A 2-D rectangular waveguide

#### Cutoff Frequencies

Any waveguide system will have cutoff frequencies: frequencies below which no waves will propagate in the guide. These determine operating bandwidth for a particular mode, or alternatively the number of simultaneous modes for a propagating frequency.

**Note:** For more information on cutoff frequencies, see Background for **2.1 1-D Waveguides: Striplines**.

## Valid Modes

When calculating cutoff frequencies, it's important to constrain the mode integers **m** and **n** appropriately. In some cases setting these integers to zero means that there is no electric field.

$$\mathbf{TM}_{mn}: E_z = E_0 \cdot \sin\left(\frac{m \cdot \pi \cdot x}{a}\right) \cdot \sin\left(\frac{n \cdot \pi \cdot y}{b}\right) \cdot e^{-j \cdot k_z \cdot z}$$

$$E_x = \frac{-j \cdot k_z \cdot k_x \cdot E_0}{\omega^2 \cdot \mu \cdot \varepsilon - k_z^2} \cdot \cos\left(\frac{m \cdot \pi \cdot x}{a}\right) \cdot \sin\left(\frac{n \cdot \pi \cdot y}{b}\right) \cdot e^{-j \cdot k_z \cdot z}$$

etc...

$$\mathbf{TE}_{mn}: E_y = E_0 \cdot \sin\left(\frac{m \cdot \pi \cdot x}{a}\right) \cdot e^{-j \cdot k_z \cdot z}$$

For TE<sub>mn</sub> modes, **n** can be zero. But the y-dependence of the TM<sub>mn</sub> modes means that **m** and **n** must both be greater than or equal to one for there to be a propagating wave in the guide.

## Mathcad Implementation

The equations that follow define propagation constants and cutoff frequencies for a rectangular waveguide.

$$\varepsilon_0 := 8.854 \cdot 10^{-12} \frac{\text{F}}{\text{m}} \quad \mu_0 := 4 \cdot \pi \cdot 10^{-7} \frac{\text{H}}{\text{m}}$$

First, enter waveguide dimensions along the x and y axes:

$$a := 3 \text{ cm} \quad b := 1.5 \text{ cm}$$

For the air-filled rectangular waveguide shown in Fig. 2.2.1, the propagation constant in the z-direction for a given propagating frequency  $\omega$  is

$$k_z(\omega, m, n) := \sqrt{\omega^2 \cdot (\mu_0 \cdot \varepsilon_0) - \left(\frac{\pi \cdot m}{a}\right)^2 - \left(\frac{\pi \cdot n}{b}\right)^2}$$

The cutoff frequency for the same guide is

$$f_c(m, n) := \frac{1}{2 \cdot \sqrt{\mu_0 \cdot \varepsilon_0}} \cdot \sqrt{\left(\frac{m}{a}\right)^2 + \left(\frac{n}{b}\right)^2}$$

Now we can enter the mode numbers, **m** and **n**, as array indices, and calculate a matrix of values for the cutoff frequency and the propagation constant, yielding design numbers for waveguides.

Enter the range of modes for **m** and **n**.

$$m := 0..4 \quad n := 0..4$$

The table below gives cutoff frequencies in GHz. The rows correspond to  $n$  from 0 to 4, and the columns to  $m$  from 0 to 4. The first row and column apply only to the  $TE_{mn}$  modes, and  $m = n = 0$  is not valid for any mode.

$$F_{n,m} := f_c(m,n)$$

	$\longleftrightarrow$ <b>m</b> $\longrightarrow$						
$F =$	0	5	9.99	14.99	19.99	GHz	$\updownarrow$ n
	9.99	11.17	14.13	18.02	22.35		
	19.99	20.6	22.35	24.98	28.27		
	29.98	30.39	31.6	33.52	36.03		
	39.97	40.28	41.2	42.69	44.69		

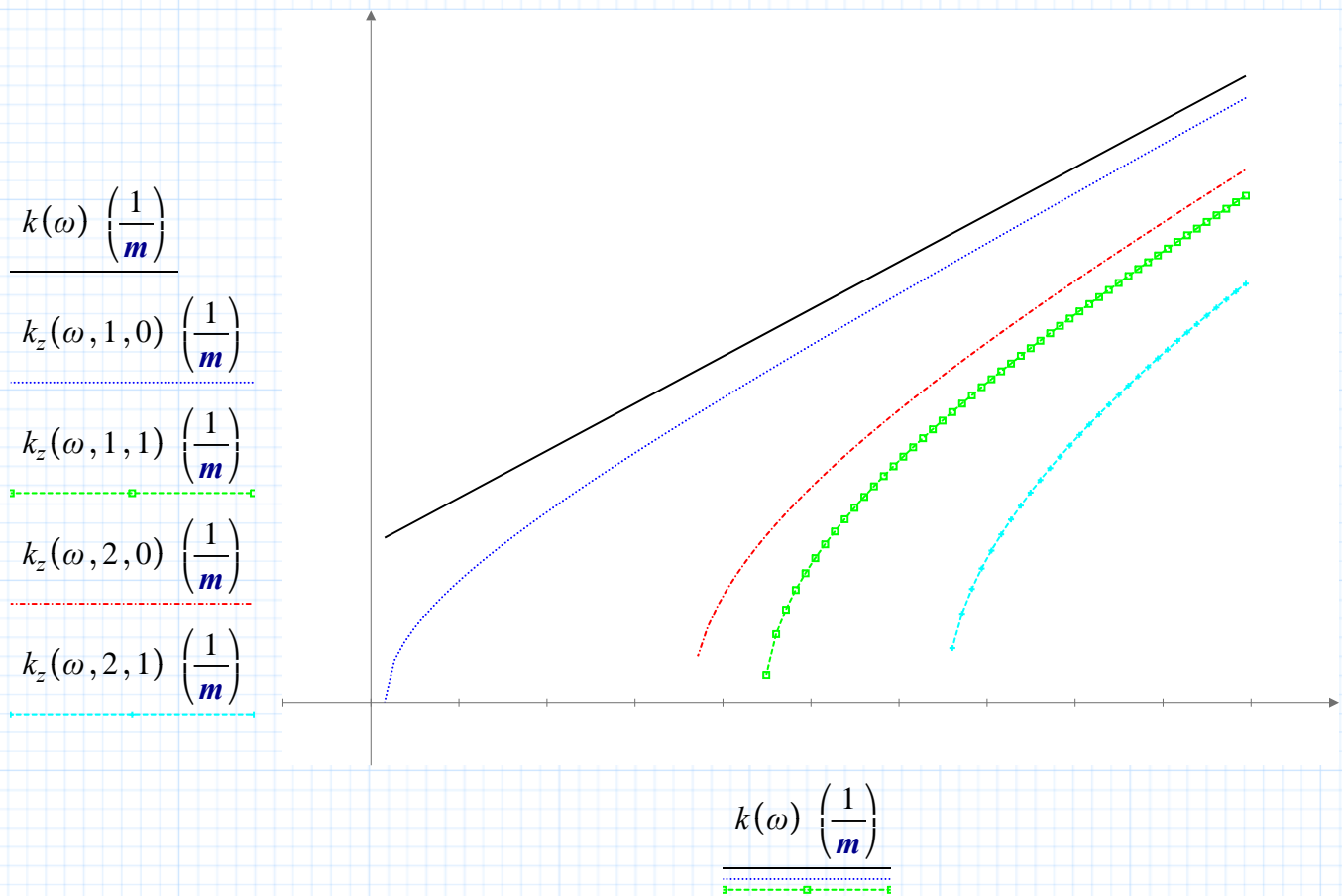
**Note:** Remember that  $m, n = 0$  is not valid for TM modes.

Using the expression for  $k_z$  and the dispersion relation, it is possible to plot the  $k_z$ - $k$  diagram, defining bandwidths for particular modes.

Begin range for frequency at the lowest cutoff frequency:

$$\omega_0 := 2 \cdot \pi \cdot f_c(1,0) \quad \omega_0 = 31.39 \text{ GHz}$$

$$k(\omega) := \omega \cdot \sqrt{\mu_0 \cdot \epsilon_0} \quad \omega := \omega_0, \omega_0 + 1 \text{ GHz} \dots 120 \text{ GHz}$$



**Fig. 2.2.2** Design curve: graph of  $k$  vs.  $k_z$ .

The x-intercepts of this diagram are the cutoff wavenumbers, and are given by the expression

$$k_c(m, n) := \sqrt{(2 \cdot \pi \cdot f_c(m, n))^2 \cdot (\mu_0 \cdot \epsilon_0)}$$

For the **a** and **b** given above, the cutoff wavenumbers are as follows:

$$k_c(1, 0) = 104.72 \frac{1}{m}$$

$$k_c(2, 1) = 296.19 \frac{1}{m}$$

Additionally, the guide wavelength  $\lambda$  is defined as a function of **kz**. The example given calculates the guide wavelength for a particular choice of dimensions, mode, and frequency.

The guide wavelength for a given frequency  $\omega$  is

$$\lambda_{guide}(\omega, m, n) := \frac{2 \cdot \pi}{k_z(\omega, m, n)}$$

For example, for the **TE<sub>10</sub>** mode propagating at three times the cutoff angular frequency, the guide wavelength is

$$\lambda_{guide}(3 \cdot \omega_0, 1, 0) = 2.12 \text{ cm}$$